

電子情報通信特論

Advanced Course of Electronics, Information and Communications

Propagation Mechanisms

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自然科学研究科 電気情報工学専攻

Propagation Mechanism

■ Reflection

- occurs when the electromagnetic wave impinges on an object which has very large dimensions as compared to the wavelength

■ Diffraction

- occurs when the radio path between the transmitter and receiver is obstructed by a surface that has sharp irregularities (edges)

■ Scattering

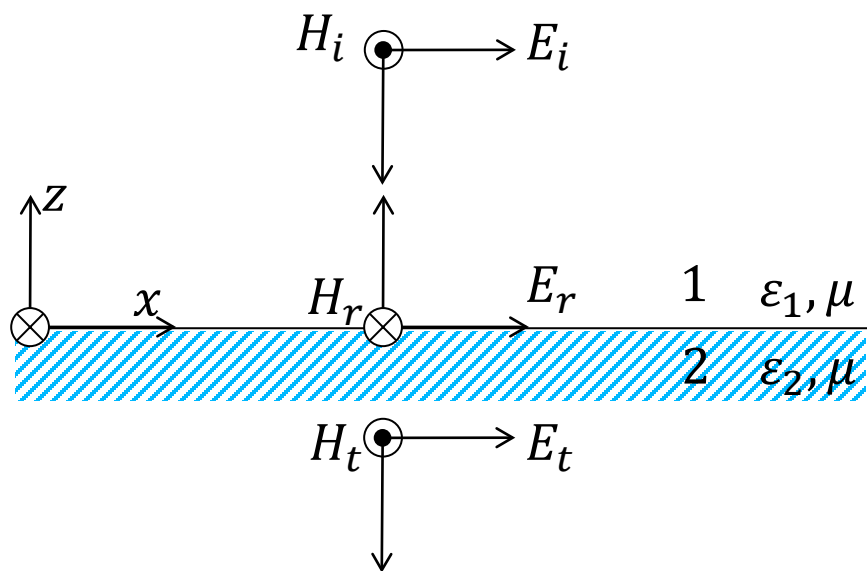
- occurs when the medium has objects that are smaller or comparable to the wavelength (small objects, rough surfaces and other irregularities)

Reflection

- Radio wave that propagates in one medium impinges upon another medium having different electric properties
- If the radio wave is incident on a dielectric material
 - Part of the energy is reflected back
 - Part of the energy is transmitted
- If the radio wave is incident on a perfect conductor
 - ??

Normal Incidence

- Consider a planar interface, and a uniform plane wave with normal incidence on the interface
- The total fields at the interface must satisfy the boundary conditions for electromagnetic fields



R : reflection coefficient
 T : transmission coefficient

E-field (x)

$$E_i = E_0 e^{+jk_1 z}$$

$$E_r = R E_0 e^{-jk_1 z}$$

$$E_t = T E_0 e^{+jk_2 z}$$

H-field (y)

$$H_i = -\frac{1}{\eta_1} E_0 e^{+jk_1 z}$$

$$H_r = \frac{R}{\eta_1} E_0 e^{-jk_1 z}$$

$$H_t = -\frac{T}{\eta_2} E_0 e^{+jk_2 z}$$

$$\mathbf{H} = \frac{1}{\eta} \hat{\mathbf{k}} \times \mathbf{E}$$

$\hat{\mathbf{k}}$: traveling direction unit vector

Normal Incidence

■ Boundary conditions ($z = 0$)

$$\hat{n} \times (\mathbf{E}_1 - \mathbf{E}_2) = 0$$

$$E_0 + RE_0 = TE_0$$

$$\hat{n} \times (\mathbf{H}_1 - \mathbf{H}_2) = 0$$

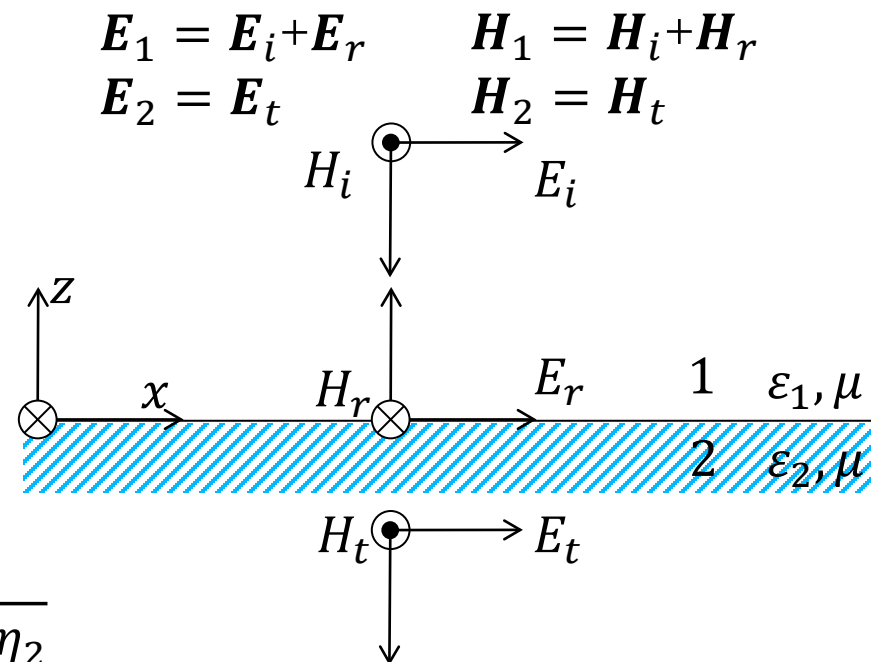
$$\frac{1}{\eta_1} E_0 - \frac{R}{\eta_1} E_0 = \frac{T}{\eta_2} E_0$$

$$R = \frac{\eta_2 - \eta_1}{\eta_1 + \eta_2}$$

Reflection coefficient

$$T = \frac{2\eta_2}{\eta_1 + \eta_2}$$

Transmission coefficient

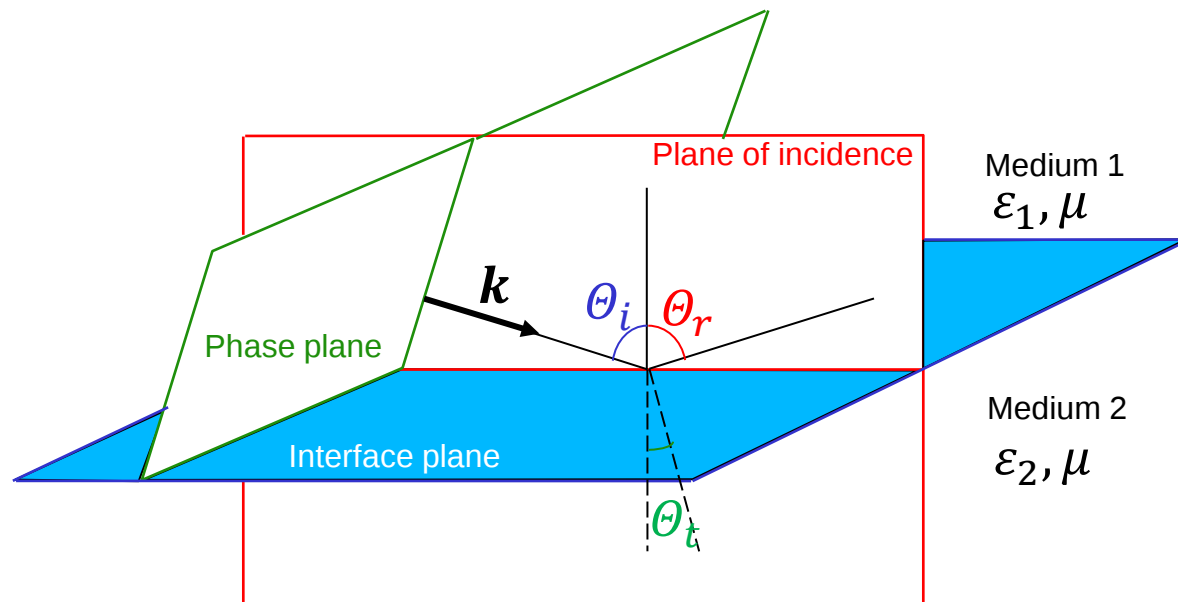


■ Special cases

- Matched medium ($\eta_2 = \eta_1$) $\rightarrow R = ?$, $T = ?$
- Perfect conductor ($\eta_2 = 0$) $\rightarrow R = ?$, $T = ?$

Oblique Incidence

- Consider a planar interface between two dielectric media. A plane wave is incident at an angle from medium 1
 - The interface plane defines the boundary between the media
 - The plane of incidence contains the propagation vector and is both perpendicular to the interface plane and to the phase planes of the wave

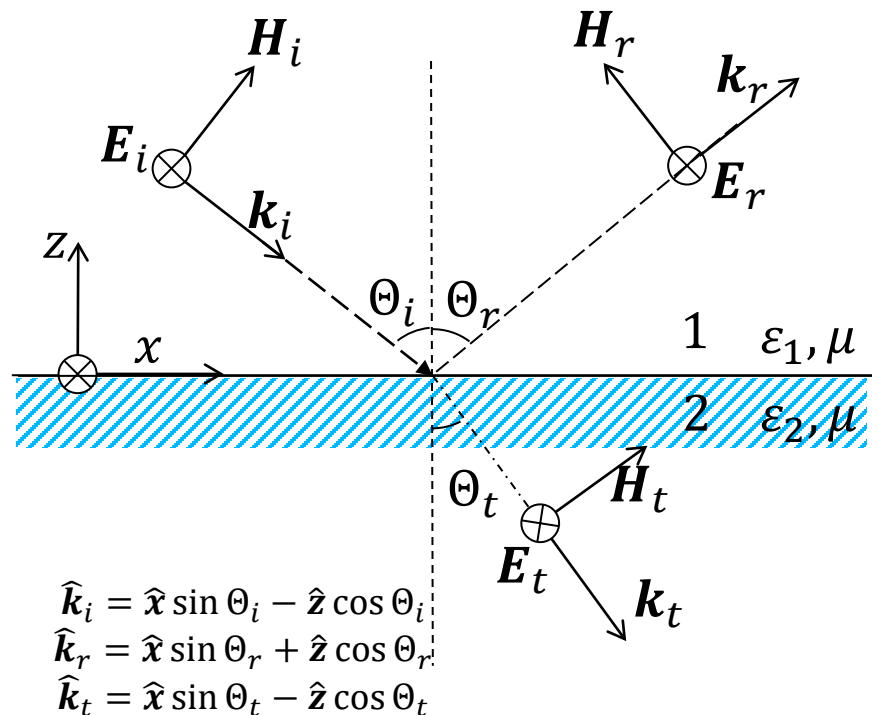


Oblique Incidence

- There are two elementary orientations (polarizations) for the electromagnetic fields:
- Perpendicular polarization ($\perp$)
 - The electric field is perpendicular to the plane of incidence and the magnetic field is parallel to the plane of incidence
 - The fields are configured as in the **Transverse Electric (TE)** modes
- Parallel polarization ($\parallel$)
 - The magnetic field is perpendicular to the plane of incidence and the electric field is parallel to the plane of incidence
 - The fields are configured as in the **Transverse Magnetic (TM)** modes
- Any plane wave with general field orientation can be obtained by superposition of two waves with perpendicular and parallel polarization

Perpendicular Polarization

- The electric field is perpendicular to the plane of incidence and the magnetic field is parallel to the plane of incidence
- The fields are configured as in the **Transverse Electric (TE)** modes



$$\begin{aligned}
 \mathbf{E}_i &= \hat{\mathbf{y}} E_0 e^{-j\mathbf{k}_i \cdot \mathbf{r}} = \hat{\mathbf{y}} E_0 e^{-jk_1(x \sin \Theta_i - z \cos \Theta_i)} \\
 \mathbf{E}_r &= \hat{\mathbf{y}} R E_0 e^{-j\mathbf{k}_r \cdot \mathbf{r}} = \hat{\mathbf{y}} R E_0 e^{-jk_1(x \sin \Theta_r + z \cos \Theta_r)} \\
 \mathbf{E}_t &= \hat{\mathbf{y}} T E_0 e^{-j\mathbf{k}_t \cdot \mathbf{r}} = \hat{\mathbf{y}} T E_0 e^{-jk_2(x \sin \Theta_t - z \cos \Theta_t)} \\
 \mathbf{H}_i &= \frac{1}{\eta_1} \hat{\mathbf{k}}_i \times \mathbf{E}_i \\
 &= (\hat{\mathbf{z}} \sin \Theta_i + \hat{\mathbf{x}} \cos \Theta_i) \frac{E_0}{\eta_1} e^{-jk_1(x \sin \Theta_i - z \cos \Theta_i)} \\
 \mathbf{H}_r &= \frac{1}{\eta_1} \hat{\mathbf{k}}_r \times \mathbf{E}_r \\
 &= (\hat{\mathbf{z}} \sin \Theta_r - \hat{\mathbf{x}} \cos \Theta_r) \frac{R E_0}{\eta_1} e^{-jk_1(x \sin \Theta_r + z \cos \Theta_r)} \\
 \mathbf{H}_t &= \frac{1}{\eta_2} \hat{\mathbf{k}}_t \times \mathbf{E}_t \\
 &= (\hat{\mathbf{z}} \sin \Theta_t + \hat{\mathbf{x}} \cos \Theta_t) \frac{T E_0}{\eta_2} e^{-jk_2(x \sin \Theta_t - z \cos \Theta_t)}
 \end{aligned}$$

Perpendicular Polarization

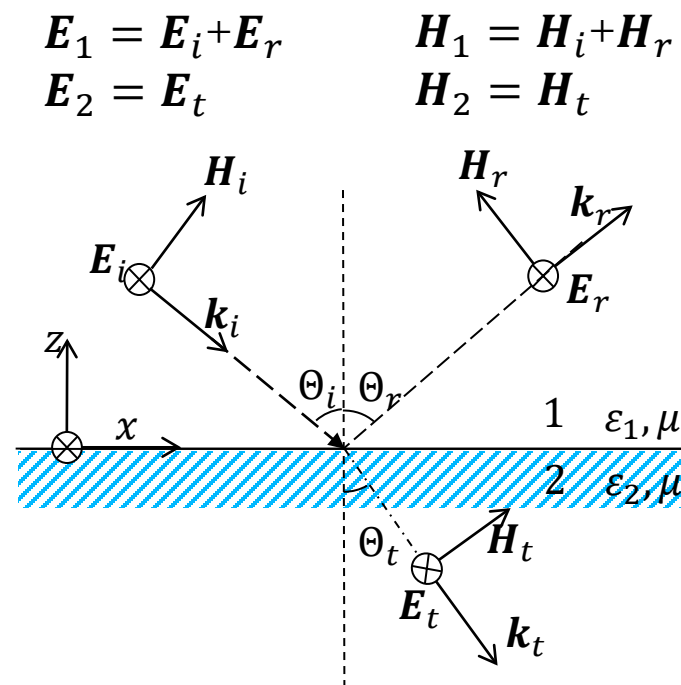
- Boundary conditions ($z = 0$)

$$\hat{n} \times (\mathbf{E}_1 - \mathbf{E}_2) = 0$$

$$e^{-jk_1 x \sin \Theta_i} + R e^{-jk_1 x \sin \Theta_r} = T e^{-jk_2 x \sin \Theta_t} \dots \dots (1)$$

$$\hat{n} \times (\mathbf{H}_1 - \mathbf{H}_2) = 0$$

$$\frac{\cos \Theta_i}{\eta_1} e^{-jk_1 x \sin \Theta_i} - \frac{R \cos \Theta_r}{\eta_1} e^{-jk_1 x \sin \Theta_r} = \frac{T \cos \Theta_t}{\eta_2} e^{-jk_2 x \sin \Theta_t} \dots \dots (2)$$



Reflection and Refraction Laws

- The relation (1) must be valid for any choice of x and we must have phase conservation law as

$$k_1 \sin \Theta_i = k_1 \sin \Theta_r = k_2 \sin \Theta_t \quad \text{where } k = \frac{2\pi}{\lambda} = \omega \sqrt{\mu \epsilon}$$

Wavenumber

- Reflection law: The angle of incidence is the same as the reflected angle

$$\Theta_i = \Theta_r$$

- Snell's law (refraction)

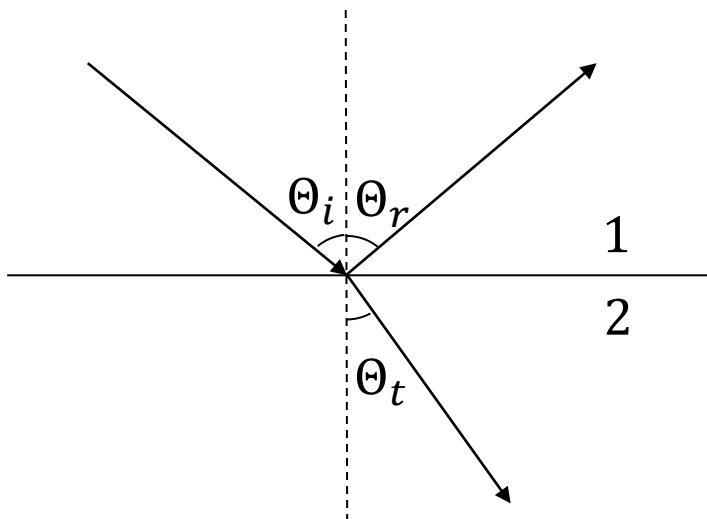
$$\sqrt{\epsilon_1} \sin \Theta_i = \sqrt{\epsilon_2} \sin \Theta_t$$

$$\frac{\sin \Theta_t}{\sin \Theta_i} = \frac{\sqrt{\epsilon_1}}{\sqrt{\epsilon_2}} = \frac{n_1}{n_2} = \frac{v_2}{v_1}$$

$$v = \frac{1}{\sqrt{\epsilon \mu}} = \frac{1}{\sqrt{\epsilon_r \mu_r}} \frac{1}{\sqrt{\epsilon_0 \mu_0}} = \frac{c}{n}$$

$$n = \sqrt{\epsilon_r} \quad (\mu_r = 1)$$

Refractive index
(屈折率)



Perpendicular Polarization

- The relations (1) and (2), we obtain

$$\left\{ \begin{array}{l} 1 + R = T \\ \frac{\cos \Theta_i}{\eta_1} - \frac{R \cos \Theta_r}{\eta_1} = \frac{T \cos \Theta_t}{\eta_2} \end{array} \right.$$

Reflection coefficient

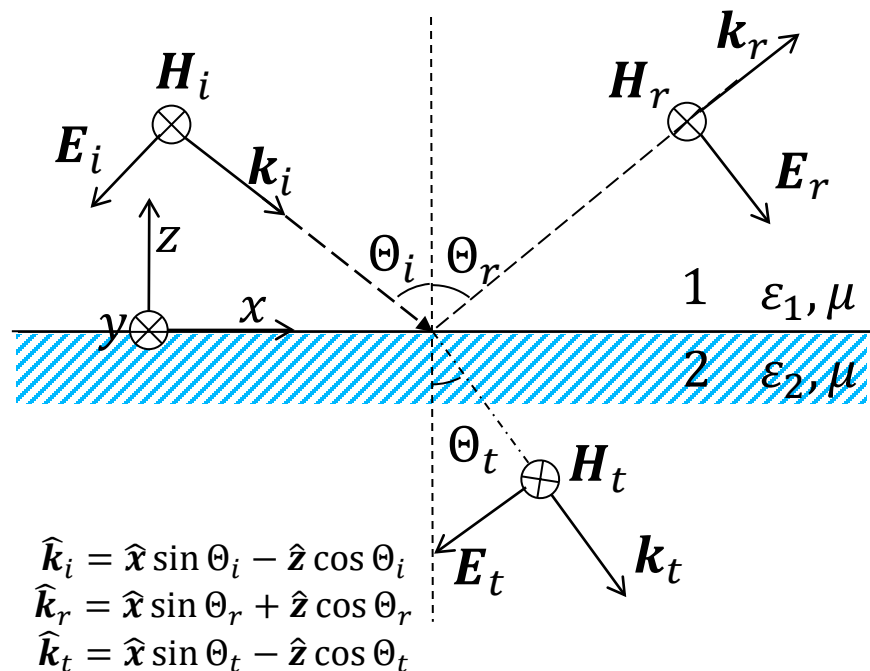
$$R_{\perp} = \frac{\eta_2 \cos \Theta_i - \eta_1 \cos \Theta_t}{\eta_1 \cos \Theta_t + \eta_2 \cos \Theta_i} = -\frac{\sin(\Theta_i - \Theta_t)}{\sin(\Theta_i + \Theta_t)}$$

Transmission coefficient

$$T_{\perp} = \frac{2\eta_2 \cos \Theta_i}{\eta_1 \cos \Theta_t + \eta_2 \cos \Theta_i} = \frac{2 \sin \Theta_t \cos \Theta_i}{\sin(\Theta_i + \Theta_t)}$$

Parallel Polarization

- The magnetic field is perpendicular to the plane of incidence and the electric field is parallel to the plane of incidence
- The fields are configured as in the **Transverse Magnetic (TM)** modes



$$\begin{aligned}
 \mathbf{H}_i &= \hat{\mathbf{y}} \frac{E_0}{\eta_1} e^{-j\mathbf{k}_i \cdot \mathbf{r}} = \hat{\mathbf{y}} \frac{E_0}{\eta_1} e^{-jk_1(x \sin \Theta_i - z \cos \Theta_i)} \\
 \mathbf{H}_r &= \hat{\mathbf{y}} \frac{RE_0}{\eta_1} e^{-j\mathbf{k}_r \cdot \mathbf{r}} = \hat{\mathbf{y}} \frac{RE_0}{\eta_1} e^{-jk_1(x \sin \Theta_r + z \cos \Theta_r)} \\
 \mathbf{H}_t &= \hat{\mathbf{y}} \frac{TE_0}{\eta_1} e^{-j\mathbf{k}_t \cdot \mathbf{r}} = \hat{\mathbf{y}} \frac{TE_0}{\eta_1} e^{-jk_2(x \sin \Theta_t - z \cos \Theta_t)} \\
 \mathbf{E}_i &= \eta_1 \mathbf{H}_i \times \hat{\mathbf{k}}_i \\
 &= (-\hat{\mathbf{z}} \sin \Theta_i - \hat{\mathbf{x}} \cos \Theta_i) E_0 e^{-jk_1(x \sin \Theta_i - z \cos \Theta_i)} \\
 \mathbf{E}_r &= \eta_1 \mathbf{H}_r \times \hat{\mathbf{k}}_r \\
 &= (-\hat{\mathbf{z}} \sin \Theta_r + \hat{\mathbf{x}} \cos \Theta_r) RE_0 e^{-jk_1(x \sin \Theta_r + z \cos \Theta_r)} \\
 \mathbf{E}_t &= \eta_2 \mathbf{H}_t \times \hat{\mathbf{k}}_t \\
 &= (-\hat{\mathbf{z}} \sin \Theta_t - \hat{\mathbf{x}} \cos \Theta_t) TE_0 e^{-jk_2(x \sin \Theta_t - z \cos \Theta_t)}
 \end{aligned}$$

Parallel Polarization

- Boundary conditions ($z = 0$)

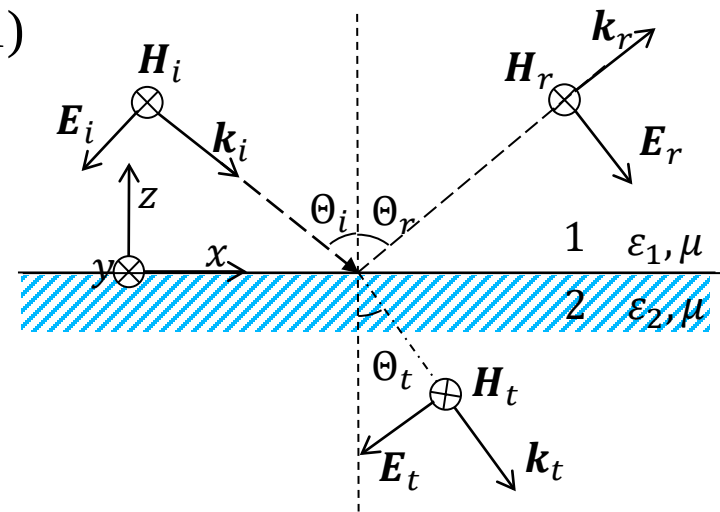
$$\hat{n} \times (\mathbf{E}_1 - \mathbf{E}_2) = 0$$

$$-\cos \Theta_i e^{-jk_1 x \sin \Theta_i} + R \cos \Theta_r e^{-jk_1 x \sin \Theta_r} = -T \cos \Theta_t e^{-jk_2 x \sin \Theta_t} \dots \dots (1)$$

$$\hat{n} \times (\mathbf{H}_1 - \mathbf{H}_2) = 0$$

$$\frac{1}{\eta_1} e^{-jk_1 x \sin \Theta_i} + \frac{R}{\eta_1} e^{-jk_1 x \sin \Theta_r} = \frac{T}{\eta_2} e^{-jk_2 x \sin \Theta_t} \dots \dots (2)$$

$$\begin{aligned} \mathbf{E}_1 &= \mathbf{E}_i + \mathbf{E}_r & \mathbf{H}_1 &= \mathbf{H}_i + \mathbf{H}_r \\ \mathbf{E}_2 &= \mathbf{E}_t & \mathbf{H}_2 &= \mathbf{H}_t \end{aligned}$$



Parallel Polarization

- Similar to perpendicular polarization

$$\left\{ \begin{array}{l} -\cos \Theta_i + R \cos \Theta_r = -T \cos \Theta_t \\ \frac{1}{\eta_1} + \frac{R}{\eta_1} = \frac{T}{\eta_2} \end{array} \right.$$

Reflection coefficient

$$R_{\parallel} = \frac{\eta_1 \cos \Theta_i - \eta_2 \cos \Theta_t}{\eta_1 \cos \Theta_i + \eta_2 \cos \Theta_t} = -\frac{\tan(\Theta_i - \Theta_t)}{\tan(\Theta_i + \Theta_t)}$$

Transmission coefficient

$$T_{\parallel} = \frac{2\eta_2 \cos \Theta_i}{\eta_1 \cos \Theta_i + \eta_2 \cos \Theta_t} = \frac{2 \sin \Theta_t \cos \Theta_i}{\sin(\Theta_i + \Theta_t) \cos(\Theta_i - \Theta_t)}$$

Total Transmission

- The reflection coefficients vanish at **Brewster angle** only in parallel polarization

$$R_{\perp} = -\frac{\sin(\Theta_i - \Theta_t)}{\sin(\Theta_i + \Theta_t)} < 0 : \text{always negative}$$

$$R_{\parallel} = -\frac{\tan(\Theta_i - \Theta_t)}{\tan(\Theta_i + \Theta_t)} = 0 \quad \text{when } (\Theta_i + \Theta_t) = 90^\circ$$

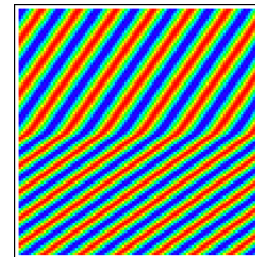
Brewster angle

$$\Theta_{i,B} + \Theta_t = 90^\circ$$

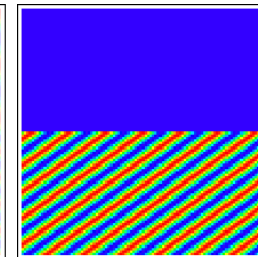
$$\frac{\sin \Theta_i}{\sin \Theta_t} = \frac{\sin \Theta_{i,B}}{\sin(90 - \Theta_{i,B})} = \frac{\sin \Theta_{i,B}}{\cos \Theta_{i,B}} = \tan \Theta_{i,B} = \frac{\sqrt{\epsilon_2}}{\sqrt{\epsilon_1}}$$

$$\Theta_{i,B} = \tan^{-1} \sqrt{\frac{\epsilon_2}{\epsilon_1}}$$

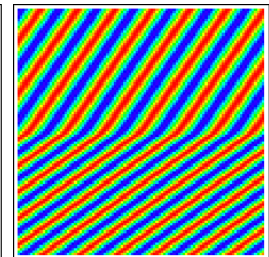
$$\epsilon_2/\epsilon_1 = 2, \quad \Theta_i = 54.7^\circ$$



Incident &
Transmitted



Reflected &
Transmitted

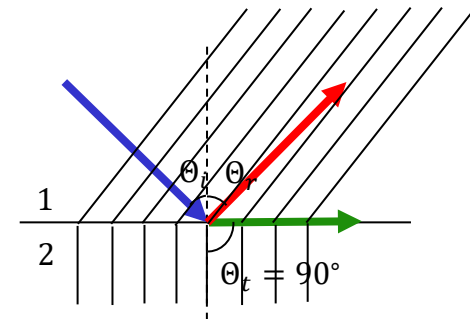


Combined &
Transmitted

Total Reflection

- We have a limit condition for total reflection, valid for both polarizations.
- When $\Theta_t = 90^\circ$, from Snell's law, we have

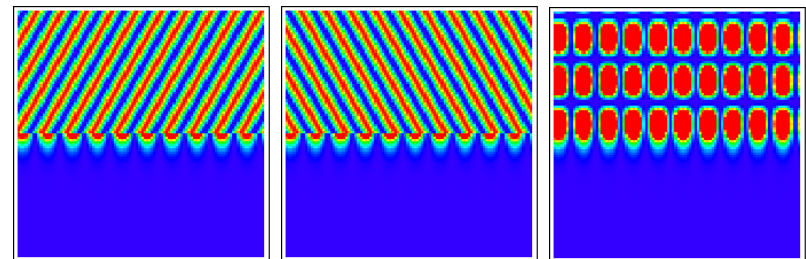
$$\frac{\sin \Theta_i}{\sin \Theta_t} = \frac{\sin \Theta_i}{1} = \frac{\sqrt{\epsilon_2}}{\sqrt{\epsilon_1}}$$



- This particular angle of incidence is called **Critical angle**

$$\Theta_{i,c} = \sin^{-1} \sqrt{\frac{\epsilon_2}{\epsilon_1}}$$

$$\epsilon_2/\epsilon_1 = 1/2, \quad \Theta_i = 60^\circ (\Theta_{i,c} = 45^\circ)$$



Incident &
Transmitted

Reflected &
Transmitted

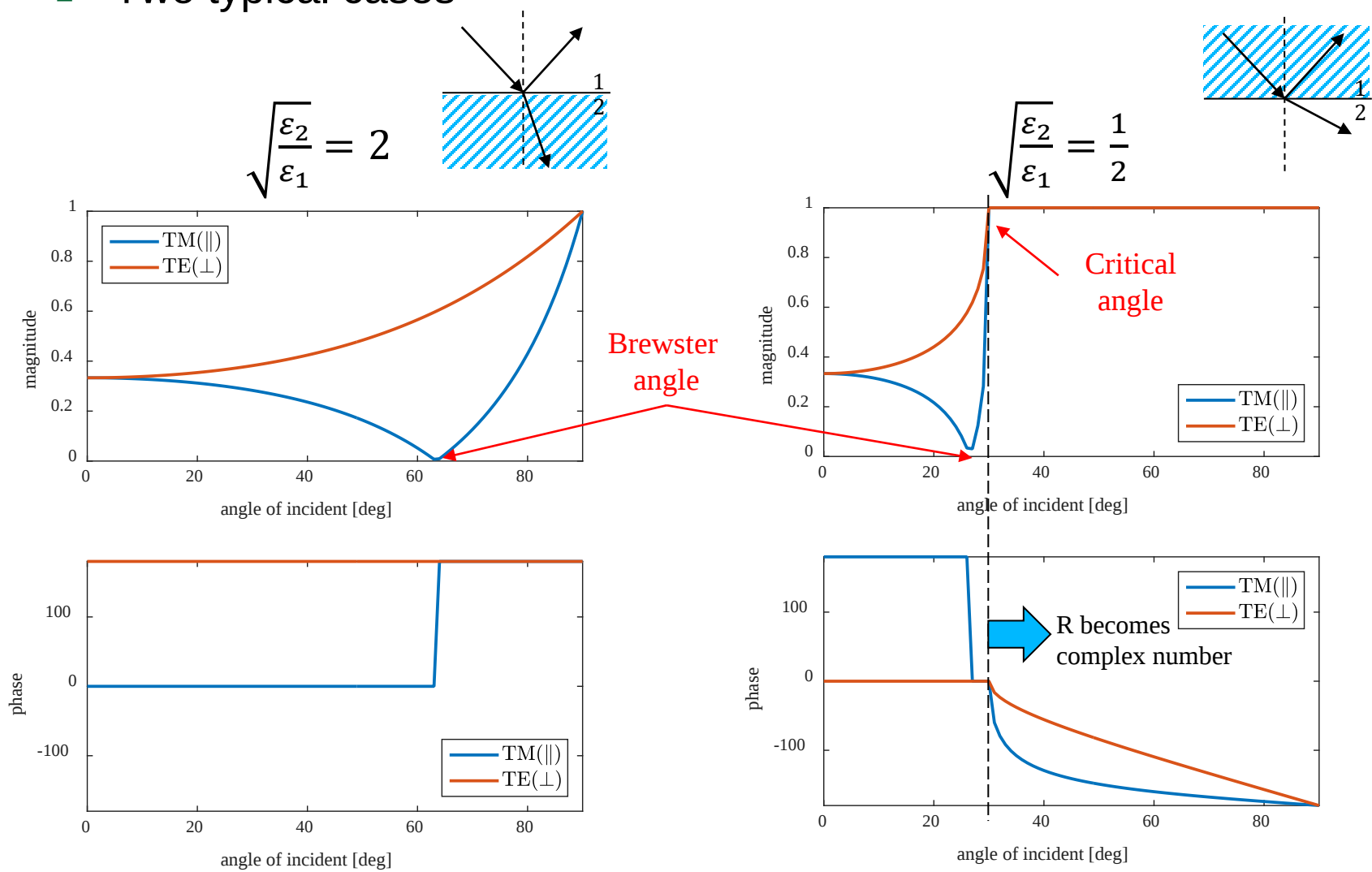
Combined &
Transmitted

Total Reflection



Reflection Coefficient

- Two typical cases



Material Constants

- Complex permittivity of interior construction materials

$$\delta = \varepsilon + \frac{\sigma}{j\omega} = \varepsilon_0 \delta_r \quad \delta_r = \varepsilon_r - j \frac{\sigma}{\omega \varepsilon_0}$$

δ_r , measured

	1 GHz	57.5 GHz	70 GHz	78.5 GHz	95.9 GHz
Concrete	7-j0.85	6.5-j0.43	—	—	6.2-j0.34
Lightweight concrete	2-j0.5	—	—	—	—
Floorboard (synthetic resin)	—	3.91-j0.33	—	3.64-j0.37	3.16-j0.39
Plaster board	—	2.25-j0.03	2.43-j0.04	2.37-j0.1	2.25-j0.06
Ceiling board (rock wool)	1.2-j0.01	1.59-j0.01	—	1.56-j0.02	1.56-j0.04
Glass	6.76-j0.09	6.76-j0.16	6.76-j0.17	6.76-j0.18	6.76-j0.19
Fibreglass	1.2-j0.1	—	—	—	—

Material Constants

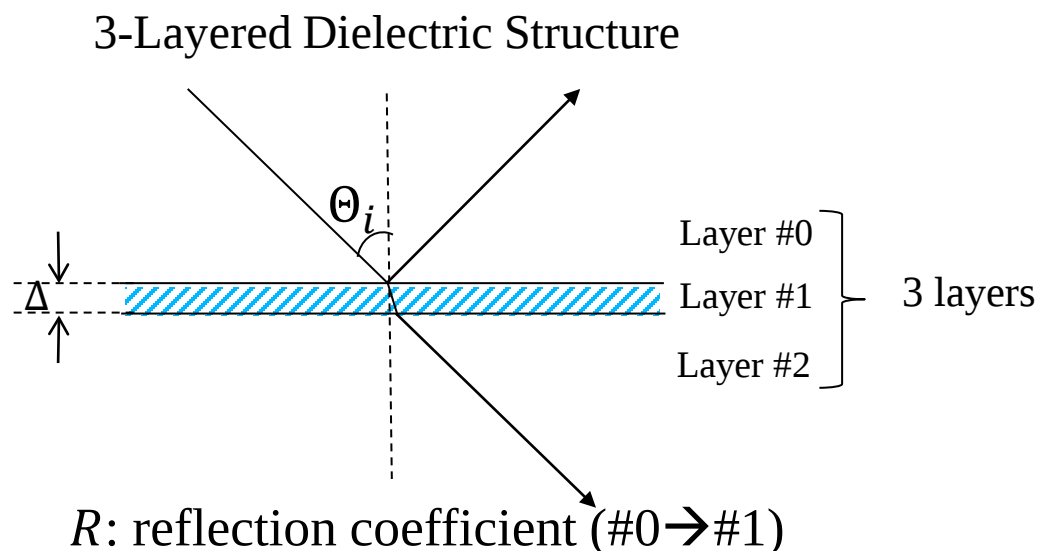
- Model parameters for the relative permittivity (ϵ_r) and conductivity (σ) of building materials

$$\sigma = c \cdot f^d \text{ where } f \text{ [GHz]}$$

Material class	Relative permittivity	Conductivity		Frequency range (GHz)
		c	d	
Concrete	5.31	0.0326	0.8095	1-100
Brick	3.75	0.038	0.0	1-10
Plasterboard	2.94	0.0116	0.7076	1-100
Wood	1.99	0.0047	1.0718	0.001-100
Glass	6.27	0.0043	1.1925	0.1-100
Ceiling board	1.50	0.0005	1.1634	1-100
Chipboard	2.58	0.0217	0.7800	1-100
Floorboard	3.66	0.0044	1.3515	50-100
Metal	1	10^7	0.0	1-100

Layered Dielectric Structures

- It occurs when a user inside a building is communicating with an outdoor BS, or in a picocell where the Mobile Station (MS) and the BS are in different rooms
- In that case, we are interested in the attenuation and phase shift of a wave transmitted through a wall
- The simplest and practically most important case occurs when a dielectric layer is surrounded on both sides by air



3-Layered Dielectric Structures

- The reflection and transmission coefficients can be determined by summation of the partial waves, resulting in a total transmission coefficient
- 3-layer case

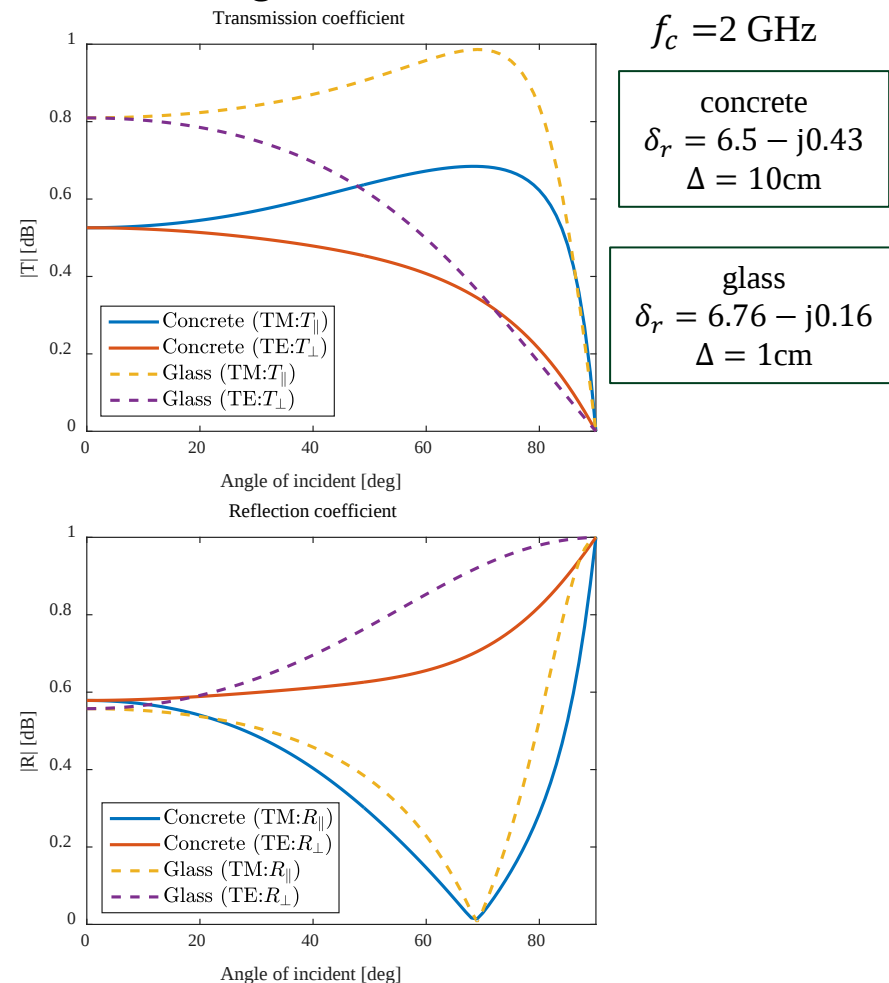
Reflection coefficient

$$R_{3\text{-layered}} = \frac{1 - e^{-j2\alpha}}{1 - R^2 e^{-j2\alpha}} R$$

Transmission coefficient

$$T_{3\text{-layered}} = \frac{(1 - R^2) e^{-j\alpha}}{1 - R^2 e^{-j2\alpha}}$$

$$\alpha = \frac{2\pi}{\lambda_0} \Delta \sqrt{\delta_r - \sin^2 \Theta_i}$$



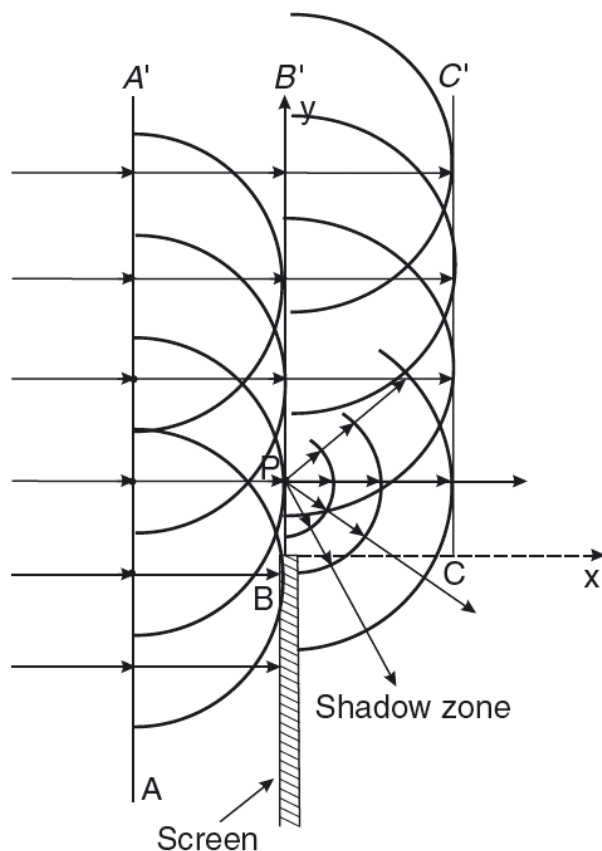
Diffraction

- When obstructed by a surface that has sharp irregularities (edges)
- Diffraction can be explained by Huygen's principle: all points on a wavefront can be considered as point sources for the production of secondary wavelets



Diffraction by Single Screen

- A homogeneous plane wave by a semi-infinite screen
- Diffraction coefficient for shadowed region



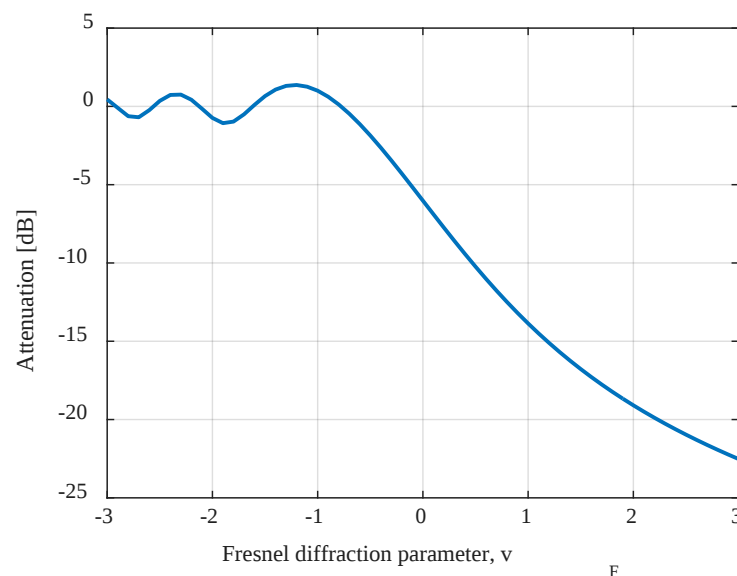
Diffraction coefficient

$$J = \left(\frac{1}{2} - \frac{1+j}{2} F(v_F) \right)$$

Fresnel Integral

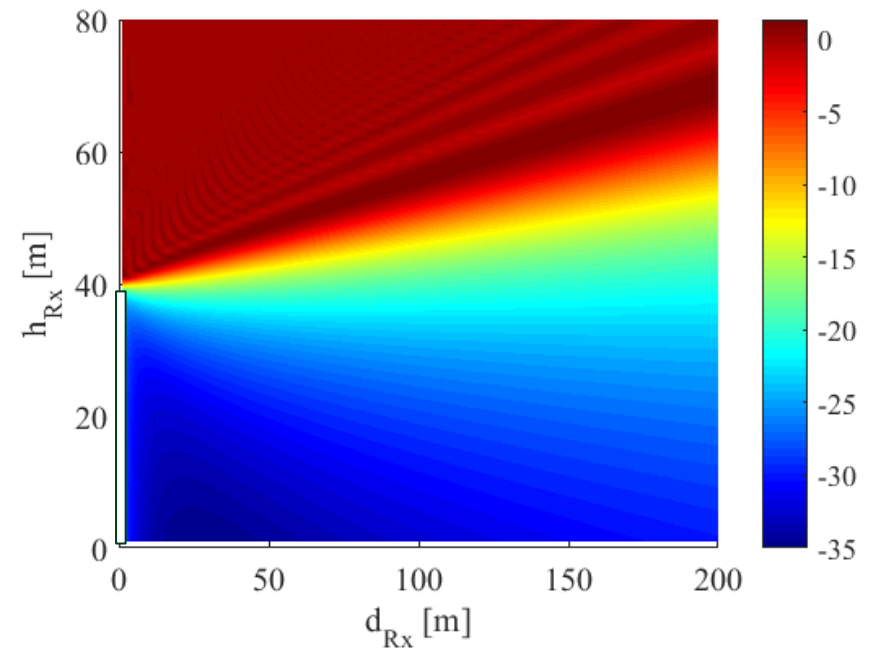
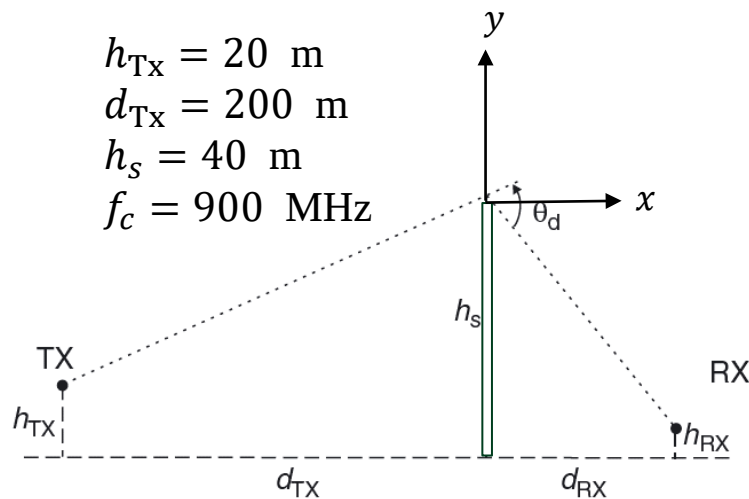
$$F(v_F) = \int_0^{v_F} \exp\left(-j\frac{\pi t^2}{2}\right) dt$$

Diffraction parameter $v_F = -2y/\sqrt{\lambda x}$



Diffraction by Single Screen

- Diffraction coefficient for shadowed region



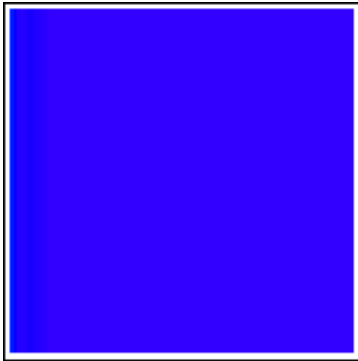
$$v_F = -2y/\sqrt{\lambda x} \approx h_s \sqrt{\frac{2(d_{TX} + d_{RX})}{\lambda d_{TX} d_{RX}}} = \theta_d \sqrt{\frac{2d_{TX} d_{RX}}{\lambda(d_{TX} + d_{RX})}}$$

Fresnel diffraction
parameter

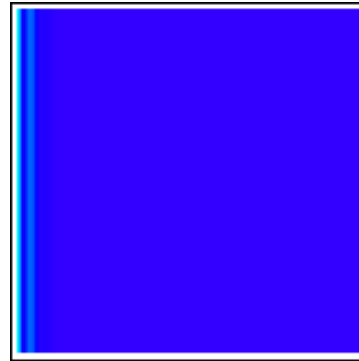
$$\theta_d = \tan^{-1} \left(\frac{h_s - h_{TX}}{d_{TX}} \right) + \tan^{-1} \left(\frac{h_s - h_{RX}}{d_{RX}} \right)$$

Examples

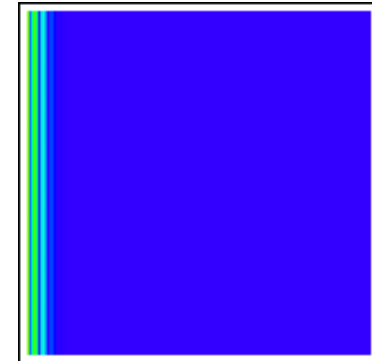
$$f = f_0$$



$$f = 2f_0$$

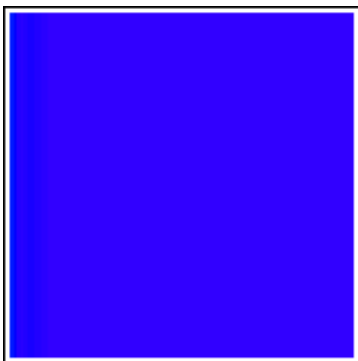


$$f = 4f_0$$

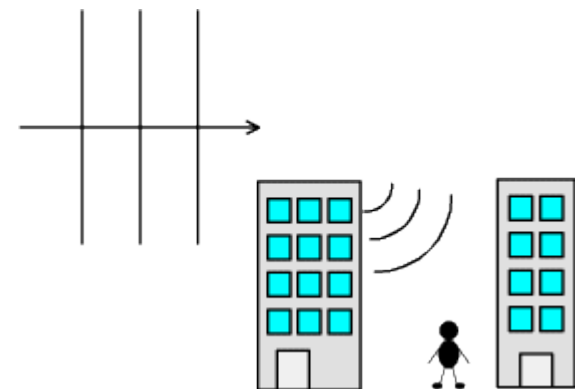
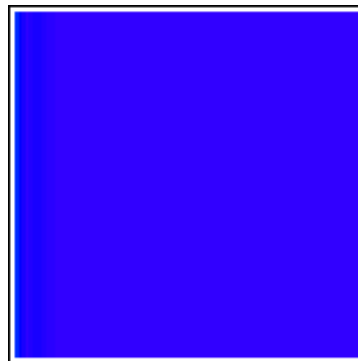


http://www.takuichi.net/hobby/edu/em/diffraction/te/lf_hf.html

TE



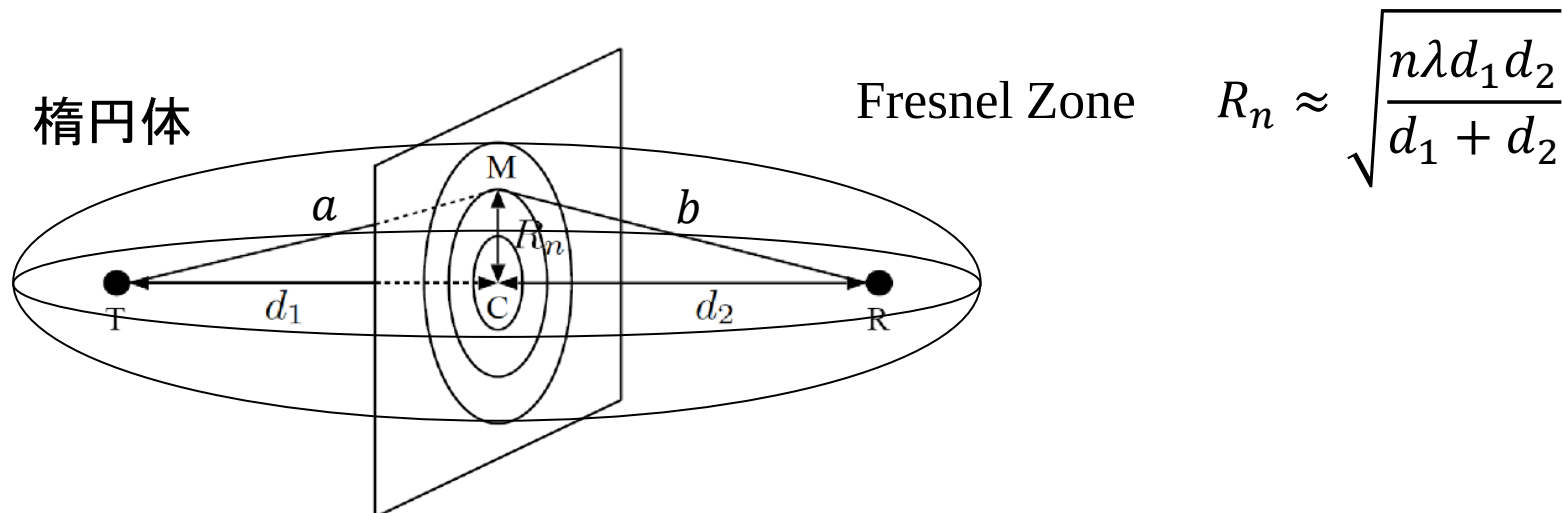
TM



<http://www.takuichi.net/hobby/edu/em/diffraction/index-j.html>

Fresnel Zones

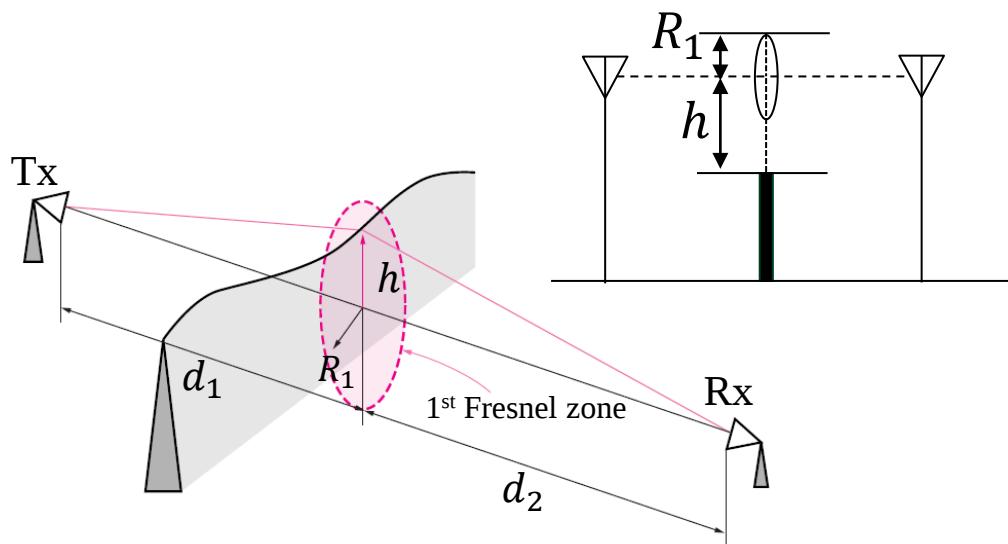
- The impact of an obstacle can also be assessed qualitatively, and intuitively, by the concept of *Fresnel zones*
- The n -th Fresnel zone is the region inside an ellipsoid defined by the locus of points where the distance $(a + b)$ is larger than the direct path between T and R ($d_1 + d_2$) by an integer multiple of $\lambda/2$
- When $d_1, d_2 \gg R_n$, Fresnel zone can be approximated by



Fresnel Zones Clearance

- Contributions within the **first zone** are all in phase, so any absorbing obstructions which do not enter this zone will have little effect on the received signal
- The Fresnel zone clearance (h/R_n) can be expressed by diffraction parameter v_F as

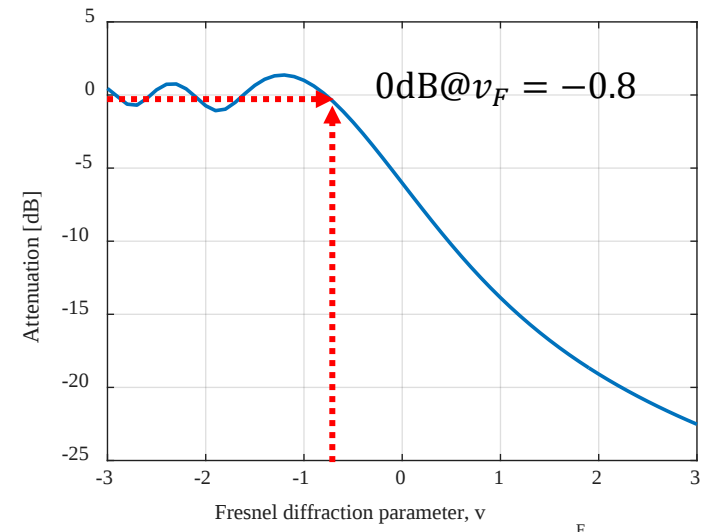
$$v_F \approx h \sqrt{\frac{2(d_1 + d_2)}{\lambda d_1 d_2}} = \frac{h}{R_1} \sqrt{2}$$



For first Fresnel zone

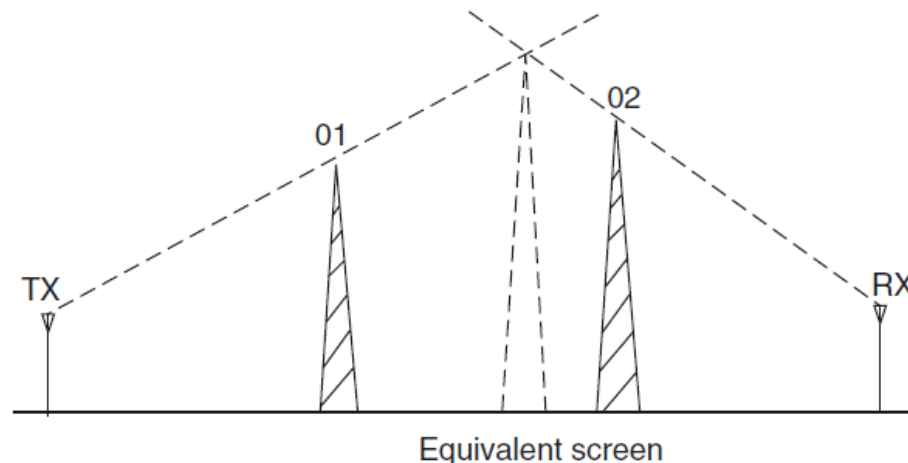
$$h/R_1 = v_F / \sqrt{2}$$

$h/R_1 \approx -0.6 \rightarrow \text{Att. 0dB}$



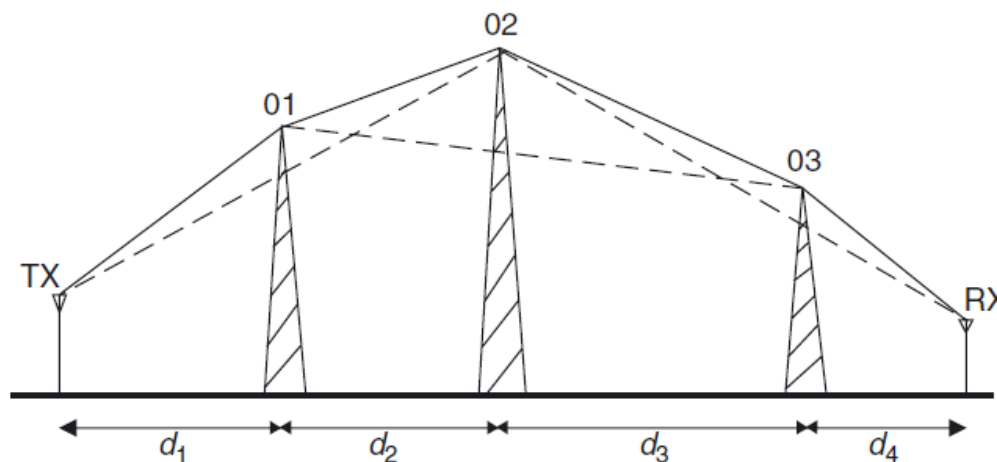
Diffraction by Multiple Screens

- The situations where *multiple* IOs are located between TX and RX are usually encountered → no exact solutions are available
- Bullington's Method
 - Replaces the multiple screens by a single, "equivalent" screen
 - The field resulting from diffraction at this single screen can be computed according to previous method → "simple"
 - Most of the physically existing screens do not impact the location of the equivalent screen → tends to give optimistic predictions of the received power



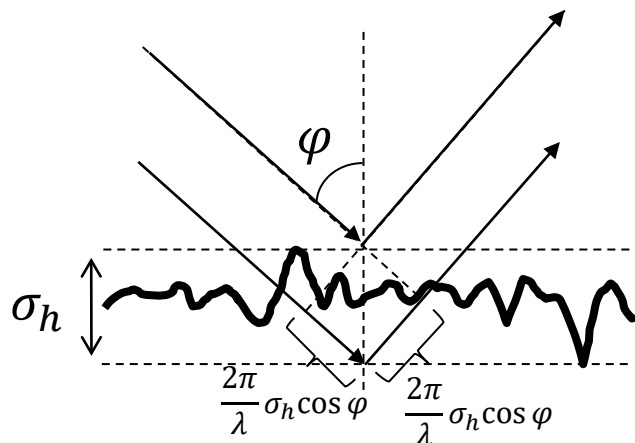
Diffraction by Multiple Screens

- The Epstein–Petersen Method
 - This approach computes the diffraction losses for each screen separately
 - Attenuations by the different screens are then added up (on a logarithmic scale) → includes the effects of all screens
 - The inaccuracies caused by this “far-field assumption” can be reduced considerably by the slope diffraction method



Scattering

- Occurs when the medium has objects that are smaller or comparable to the wavelength (small objects, rough surfaces and other irregularities on the channel)
- Causes the transmitter energy to be radiated in many direction e.g. forage, street sign, lampposts
- Surface roughness criteria



$$\Delta\phi = \frac{4\pi}{\lambda} \sigma_h \cos \varphi \quad : \text{Rayleigh roughness}$$

$$\Delta\phi > \frac{\pi}{2} \quad \Rightarrow \quad \sigma_h > \frac{\lambda}{8 \cos \varphi} \quad \text{Rayleigh criterion}$$

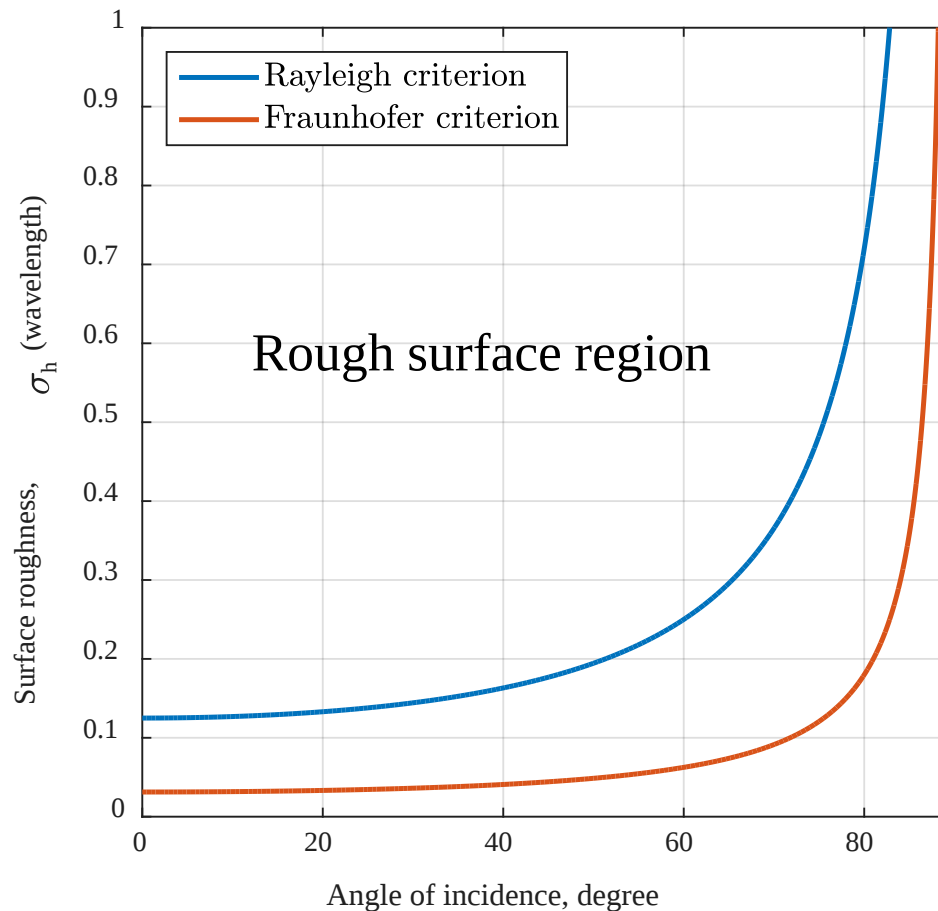
$$\Delta\phi > \frac{\pi}{8} \quad \Rightarrow \quad \sigma_h > \frac{\lambda}{32 \cos \varphi} \quad \text{Fraunhofer criterion}$$

σ_h : standard deviation of height distribution

Scattering

- Surface roughness criteria

Surfaces above the curve cannot be accurately modelled using the Fresnel reflection coefficient alone



Scattering

- Scattering coefficients by Kirchhoff theory

$$\rho_s = \exp\left(-\frac{1}{2}(\Delta\phi)^2\right) = \exp\left(-\frac{1}{2}\left(4\pi \frac{\sigma_h}{\lambda} \cos \varphi\right)^2\right)$$

